

# Unified Analytical Model of Threshold Voltage in Symmetric and Asymmetric Double-Gate MOSFETs

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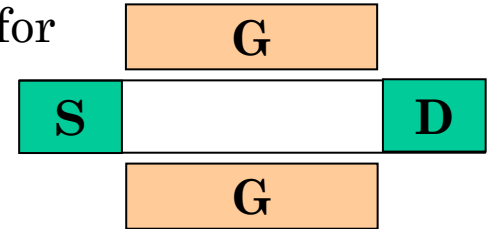


# Outline

- ◆ Introduction
- ◆ Threshold voltage model for long channels
- ◆ Model validation
  - Threshold voltage
  - Potential distribution
  - Surface potential
- ◆ Symmetric vs. asymmetric Double-Gate
  - $V_T$  requirements for ultimate integration
  - $V_T$  sensitivity to structure parameters
- ◆ Conclusion

# Introduction

- ◆ **Double-Gate:** one of the most promising structures for ultimate deca-nanometer scale



- ◆ **DG threshold voltage modeling:**
  - $V_T$  definition
  - Mobile charge usually neglected

- ◆ **Existing models:**

		Model 1	Model 2	Model 3
<b>Structure</b>	Symmetric	YES	YES	NO
	Asymmetric	YES	NO	YES
<b>Channel Doping</b>	Intrinsic	YES	YES	YES
	High	NO	YES	NO
<b>Mobile charge (in Poisson equation)</b>		YES	YES	NO

*Model 1: Y. Taur, IEEE TED 48, p. 2861, 2001.*

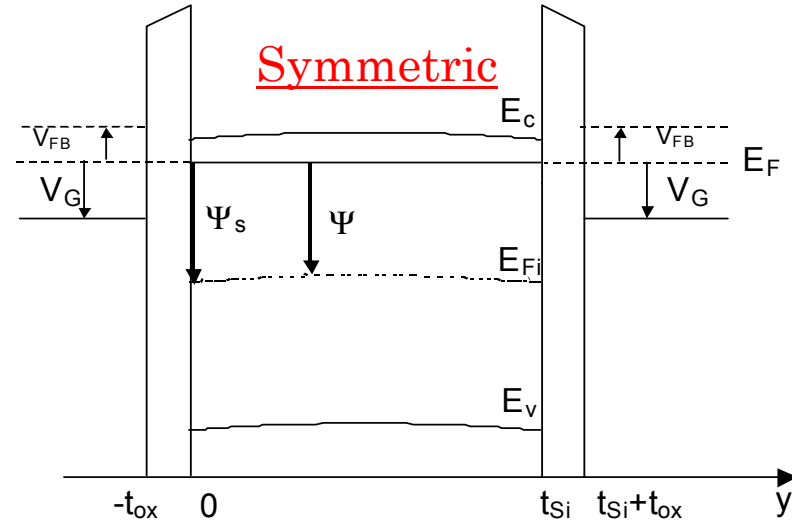
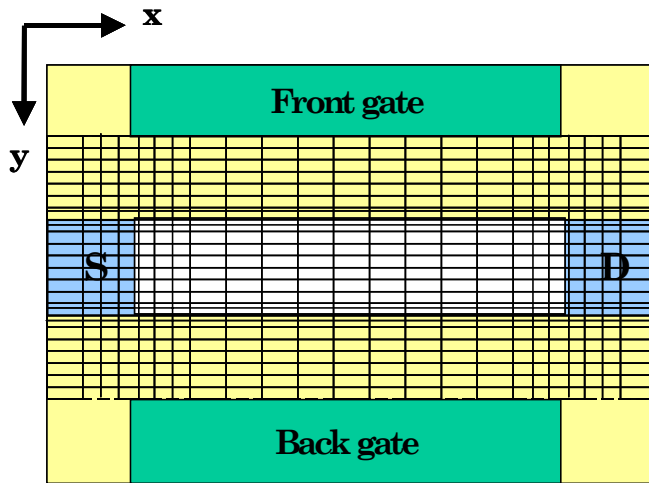
*Model 2: P. Francis et al., IEEE TEL 41, p. 715, 1994.*

*Model 3: K. Suzuki et al., IEEE TEL 42, p. 1940, 1995.*

- ◆ **This work: unified  $V_T$  model for both symmetric and asymmetric structures, with doped or undoped bodies and considering the mobile charge**

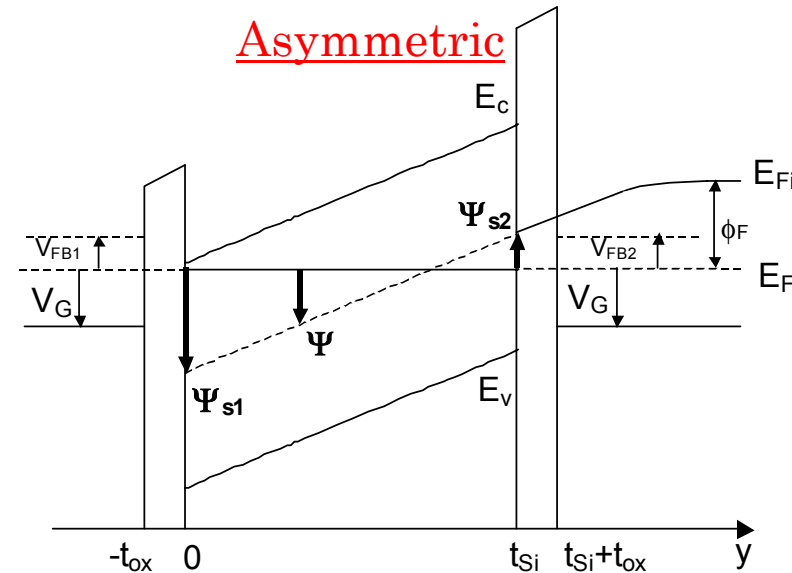
# Threshold voltage model

# Band Diagram

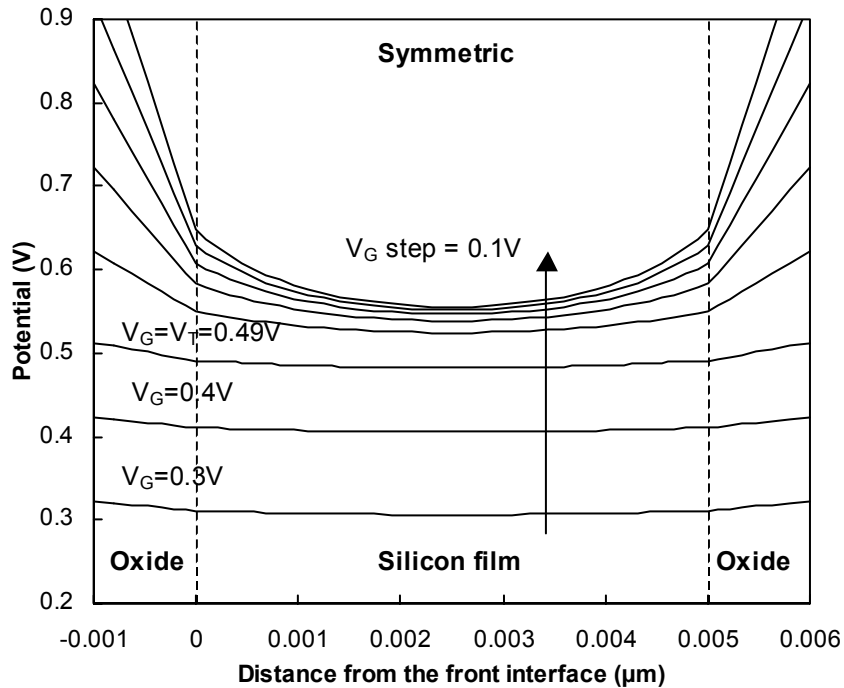


- ◆ Long channel model
- ◆ Gauss law

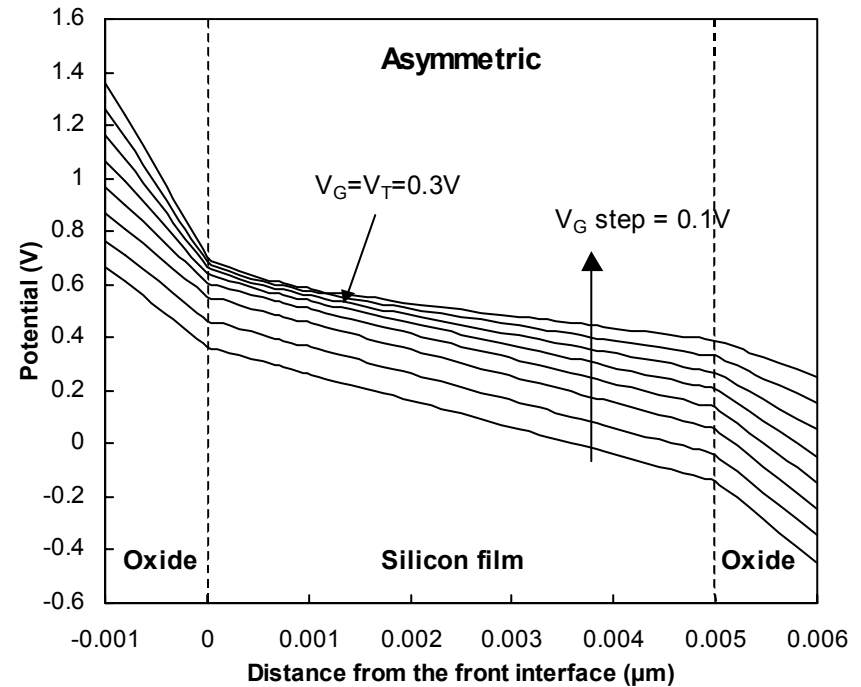
$$\begin{cases} V_G - V_{FB1} = \frac{\epsilon_{Si}}{\epsilon_{ox}} t_{ox} E_1 + \Psi_{s1} + \phi_F \\ V_G - V_{FB2} = -\frac{\epsilon_{Si}}{\epsilon_{ox}} t_{ox} E_2 + \Psi_{s2} + \phi_F \end{cases}$$



## Symmetric



## Asymmetric



- ◆ Potential approximation:

$$\Psi = \Psi_{s1} + \alpha y + \beta y^2$$

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◆ Electric field at interfaces:

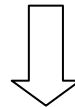
$$E_1 = -\left. \frac{d\Psi}{dy} \right|_{y=0} = -\alpha \quad E_2 = -\left. \frac{d\Psi}{dy} \right|_{y=t_{Si}} = -\alpha - 2\beta t_{Si}$$

◆ 1-D Poisson equation:

$$\underbrace{\int_0^{t_{Si}} \frac{d^2\Psi}{dy^2}}_{E_1 - E_2} = \int_0^{t_{Si}} \frac{qN_A}{\epsilon_{Si}} + \frac{1}{\epsilon_{Si}} \int_0^{t_{Si}} q n_i e^{\frac{q\Psi}{kT}}$$

$Q_i$  (inversion charge)

$$E_1 - E_2$$



$$\beta(Q_i) = \frac{qN_A}{2\epsilon_{Si}} + \frac{Q_i}{2\epsilon_{Si} t_{Si}}$$

$$\alpha(Q_i) = \frac{V_{FB1} - V_{FB2}}{2\gamma t_{ox} + t_{Si}} - \beta(Q_i) t_{Si}$$

◆ Conventional MOSFET: band bending is  $2 \times \phi_F$

◆ Double-Gate MOSFET:  $\psi_s \neq 2 \times \phi_F$

◆ **Threshold voltage definition:**

$$V_G = V_T \text{ when } \frac{d^2 I_D}{dV_G^2} \text{ is maximum} \implies \boxed{\frac{d^3 I_D}{dV_G^3} = 0}$$

◆ Drain current:

$$I_D = \frac{W}{L} \mu_n \int_0^{V_D} Q_i(V) dV$$

low  $V_D$  approximation

$$\implies \boxed{\frac{d^3 Q_i}{dV_G^3} = 0}$$

- ◆ Surface potential

$$\left. \begin{aligned} Q_i &= \int_0^{t_{Si}} q n_i e^{\frac{q\Psi(y)}{kT}} dy \\ \Psi &= \Psi_{s1} + \alpha y + \beta y^2 \end{aligned} \right\} \Rightarrow \Psi_{s1} = \frac{kT}{q} \ln\left(\frac{Q_i}{qn_i}\right) + \Psi_1$$

where  $\Psi_1 = -\frac{kT}{q} \ln \int_0^{t_{Si}} e^{\frac{q}{kT}(\alpha y + \beta y^2)} dy$

- ◆  $\Psi_1$  can be expressed analytically:

$$\Psi_1 = \frac{e^{-\frac{\alpha^2}{4u_{th}\beta}} \sqrt{\pi u_{th}} \operatorname{Erfi}\left(\frac{\alpha}{2\sqrt{u_{th}\beta}}\right)}{2\sqrt{\beta}} - \frac{e^{-\frac{\alpha^2}{4u_{th}\beta}} \sqrt{\pi u_{th}} \operatorname{Erfi}\left(\frac{\alpha + 2t_{Si}\beta}{2\sqrt{u_{th}\beta}}\right)}{2\sqrt{\beta}}$$

- ◆  $\Psi_1$  is linear as a function of  $Q_i$

$$\Psi_1 = M + N \times Q_i$$

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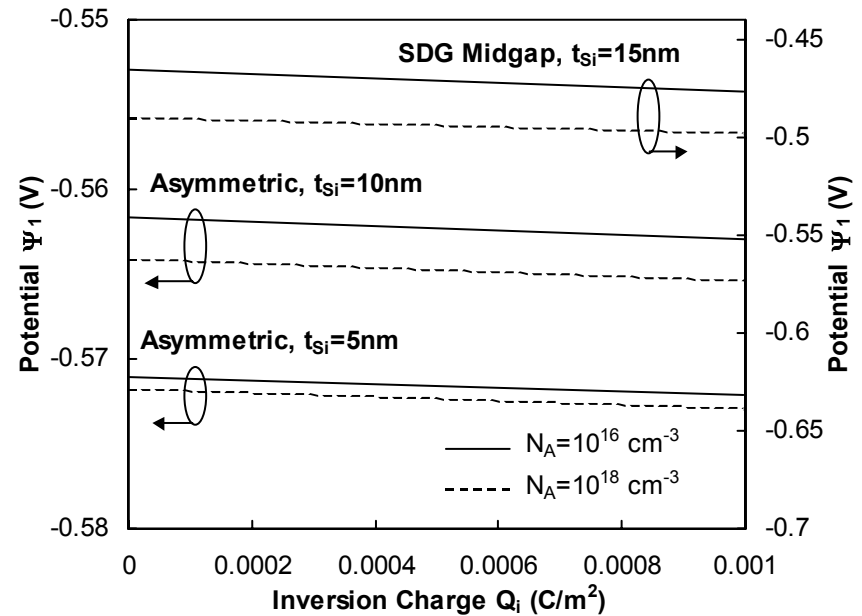
$$M = \Psi_{1\max}$$

$$N = \frac{\partial \Psi_1}{\partial Q_i} = \frac{\Delta \Psi_1}{\Delta Q_i}$$

$$\left\{ \begin{array}{l} \Psi_{1\max} = \Psi_1(Q_i = Q_{i\min}) \\ \Psi_{1\min} = \Psi_1(Q_i = Q_{i\max}) \\ Q_{i\min} = qn_i^2 \frac{t_{Si}}{N_A} \\ Q_{i\max} = (V_G - V_T)C_{ox} = \frac{kT}{q} C_{ox} \end{array} \right.$$

- ◆ Surface potential

$$\Psi_{s1} = \frac{kT}{q} \ln\left(\frac{Q_i}{qn_i}\right) + M + NQ_i$$



$$\frac{d^3 Q_i}{dV_G^3} = 0$$

$$\Psi_{s1} = \frac{kT}{q} \ln\left(\frac{Q_i}{qn_i}\right) + M + NQ_i$$

$$V_G - V_{FB1} = \frac{\epsilon_{Si}}{\epsilon_{ox}} t_{ox} E_1 + \Psi_{s1} + \phi_F$$



$$Q_{iT} = \frac{\frac{kT}{q} C_{ox}}{1 - 2N \frac{kT}{q} C_{ox}}$$

Inversion charge at threshold

◆  $\alpha$  and  $\beta$  at threshold

$$\alpha(Q_{iT}) = \frac{V_{FB1} - V_{FB2}}{2\gamma t_{ox} + t_{Si}} - \frac{qN_A t_{Si}}{2\epsilon_{Si}} + \frac{\frac{kT}{q} C_{ox}}{2\epsilon_{Si} \left(1 - 2N \frac{kT}{q} C_{ox}\right)}$$

$$\beta(Q_{iT}) = \frac{qN_A}{2\epsilon_{Si}} + \frac{\frac{kT}{q} C_{ox}}{2\epsilon_{Si} t_{Si} \left(1 - 2N \frac{kT}{q} C_{ox}\right)}$$

◆ Surface potential at threshold

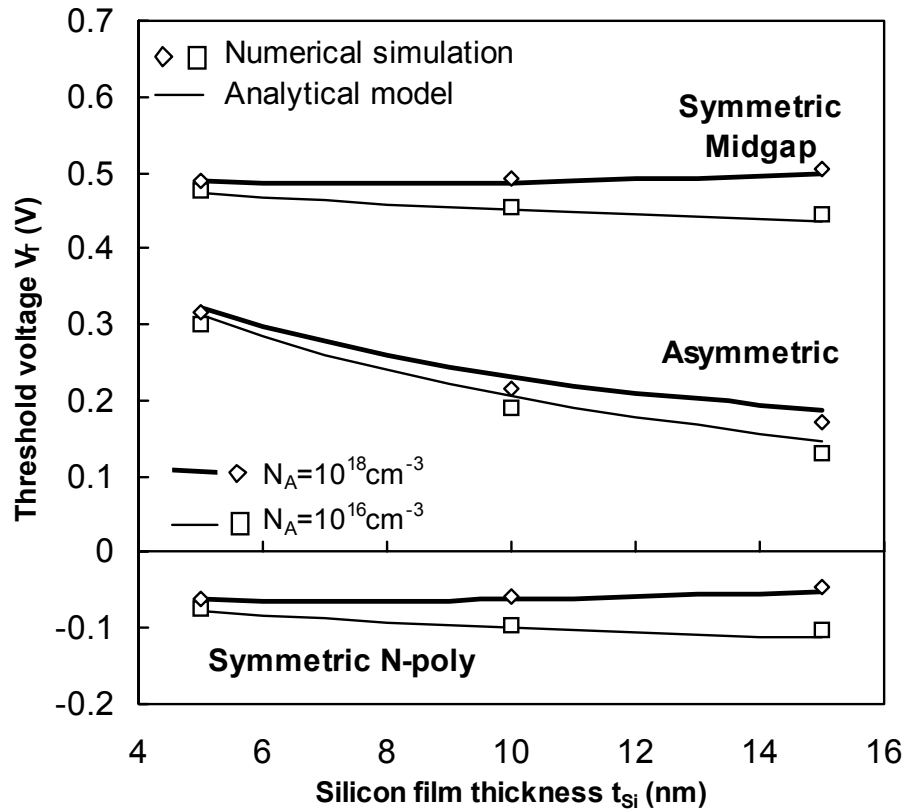
$$\Psi_{s1T} = \frac{kT}{q} \ln \left[ \frac{\frac{kT}{q} C_{ox}}{qn_i \left(1 - 2N \frac{kT}{q} C_{ox}\right)} \exp \left( M - \frac{N \frac{kT}{q} C_{ox}}{1 - 2N \frac{kT}{q} C_{ox}} \right)^{-1} \right]$$

◆ Final expression of the threshold voltage

$$V_T = V_{FB1} \frac{\gamma t_{ox} + t_{Si}}{2\gamma t_{ox} + t_{Si}} + V_{FB2} \frac{\gamma t_{ox}}{2\gamma t_{ox} + t_{Si}} + \Psi_{s1T} + \gamma t_{ox} t_{Si} \beta(Q_{iT}) + \phi_F$$

# Model validation (1)

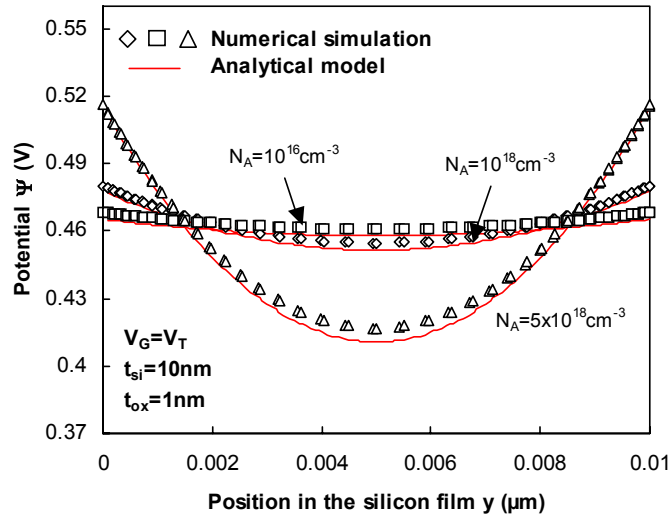
## Comparison with numerical simulation



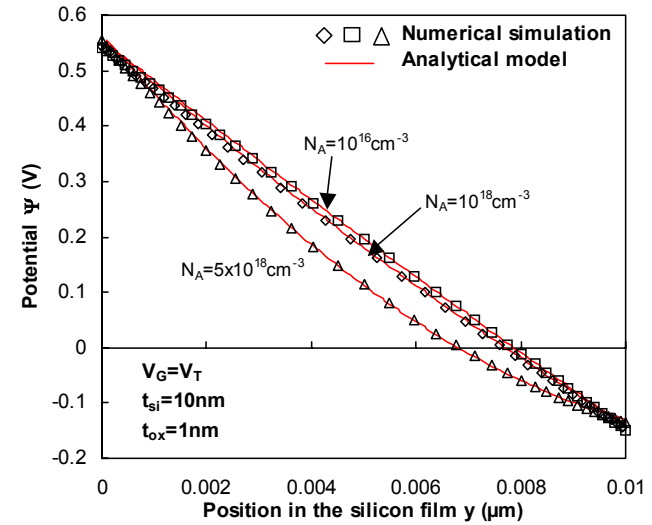
# Model validation (2)

## Potential distribution

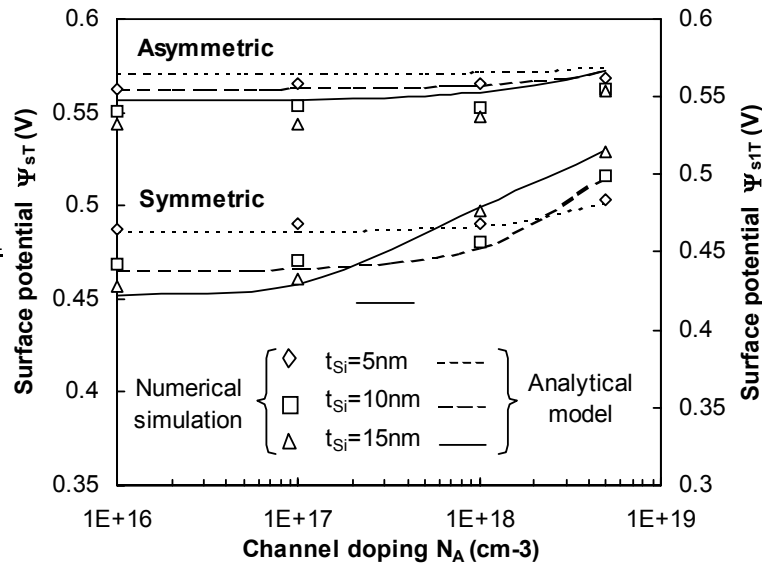
### Symmetric DG



### Asymmetric DG



## Surface potential



# Conclusion

- ❑ A unified analytical model for the threshold voltage in long channel Double-Gate devices is developed
- ❑ Compared to previous models our approach gives a unique expression for the threshold voltage applying to both symmetric and asymmetric DG structures with both doped and undoped films
- ❑ The model takes into account the mobile charge in the Poisson equation
- ❑ The model is used to analyze the threshold voltage sensitivity to the silicon thickness and the gate requirements for nano-scale integration